

Nonlinear Analysis of Wiggler Taper, Mode Competition, and Space-Charge Effects For a 280-GHz Free Electron Laser

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Abstract—The nonlinear characteristics of a 280-GHz free electron laser (FEL) are simulated by numerical computation. The three-dimensional set of coupled nonlinear differential equations is solved for a set of TE and TM modes valid for the high gain Compton regime. The use of a nonlinear taper for efficiency enhancement, the sensitivity of gain to competing mode power levels, space-charge effects, and the effect of electron beam source distributions on gain and efficiency are examined for a 10-MeV 3-kA beam. It is found that the nonlinear taper greatly enhances the gain and efficiency and makes the saturation power levels relatively insensitive to the competing mode power levels. The efficiency is increased to 48% by means of a nonlinear taper in which the 3-D and wiggler-averaged codes are compared and the effects of space charge are found to reduce the efficiency to a level of 32%. The effect of beam quality in terms of the four volume phase space is examined and found to have an observable effect at this wavelength.

I. INTRODUCTION

THE FREE ELECTRON laser has been shown [1] to operate quite effectively at high powers and efficiency in the 35-GHz microwave range. The scaling of these results to the millimeter-wave range at 280 GHz is currently underway [2], [3], and applications to the areas of communications, high resolution radar, spectroscopy, and the heating of fusion plasmas are envisioned. Work by Ganguly and Freund [4], [5] and Freund *et al.* [6], [7] has formulated the three-dimensional analysis and set of coupled nonlinear differential equations which self-consistently solve for the ensemble of electrons and mode wave fields in a circular or rectangular waveguide. Beam-self and space-charge fields are neglected in the analysis so that it is valid for the high-gain Compton regime. We previously modified this analysis and examined computational solutions for a helical quadrupole wiggler [8], [9] operating at 35 and 300 GHz. Ganguly and Freund [10] recently utilized an approximate method which expands the space-charge field in terms of the Trivelpiece-Gould modes for a fully filled cylindrical waveguide. This method is found to accurately model experiments in which the beam fills most of the guide radius and the case of

near grazing intersection between the beam line and waveguide dispersion curves.

In this paper, we examine the effect of a nonlinear wiggler taper and the sensitivity of gain to competing mode power levels, we compare the 3-D solutions to a wiggler-averaged solution including space-charge forces, and we investigate the effect of electron beam source distributions on the gain and efficiency for a 280-GHz FEL. The analysis and system of equations for the Compton regime are similar to that derived by Freund *et al.* [6], [7]. In this paper, we examine the effect of a nonlinear taper and electron source distributions on FEL gain and efficiency. For the nonlinear taper, Kroll *et al.* [11] (abbreviated as KMR henceforth) and Cover *et al.* [12] examined its effect on a single mode; we extend this to investigate its effect for the other competing modes, since a highly overmoded waveguide is usually employed for guiding radiation at millimeter wavelengths. Goldstein *et al.* [13] have examined the effect of electron beam source distribution in the optical regime and found that there is a factor of ten difference in the small-signal gain of the FEL oscillator for different distributions. Colson and Blau [14] and Boyd *et al.* [15] have examined phase space source distributions and their effect on gain. For the millimeter-wave FEL amplifier of interest, we look at effects from the small-signal to the saturation regimes.

In Section II, the set of governing equations is derived including simultaneously interacting waveguide modes, the hypervolume phase space distribution for an electron beam, and the prescription for the optimum taper rate. Section III describes the numerical solution for a 280-GHz case in which the nonlinear taper, mode competition, a comparison with a wiggler averaged code, and space charge effects are considered. The effects of a Gaussian and uniform-source phase space representation on the gain and efficiency are examined as well.

II. ANALYSIS OF THE INTERACTION DYNAMICS

The schematic configuration of an FEL amplifier with a linear dipole wiggler is shown in Fig. 1. The relativistic electron beam is described as in accelerator physics by a distribution function in phase space. Assuming that the electron beam is monoenergetic and that none of the electrons intersect the wall, the current density can be written

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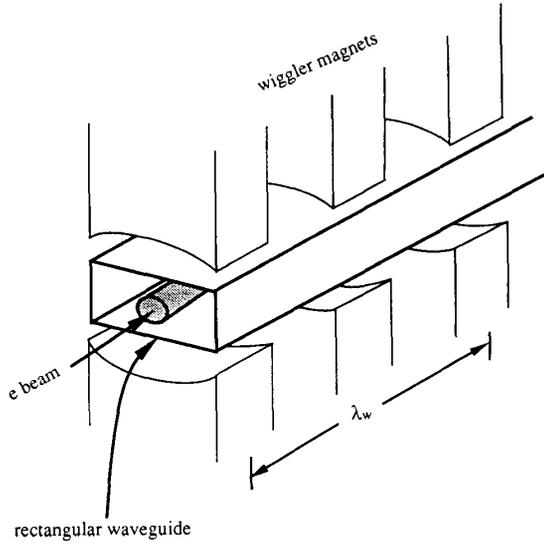


Fig. 1. Schematic view of FEL amplifier with dipole wiggler configuration.

as follows:

$$\begin{aligned}
 J(\mathbf{x}, t) = & -en_b \iiint_{\Omega_4} dp_{x0} dp_{y0} dx_o dy_o \\
 & \cdot D(\mathbf{p}'_{\perp}, \mathbf{x}'_{\perp}) v_z \int_{-T/2}^{T/2} dt_o \sigma_{\parallel}(t_o) \\
 & \times v(z; \mathbf{p}'_{\perp}, \mathbf{x}'_{\perp}, t_o) \delta^2(\mathbf{x}_{\perp} - \xi_{\perp}) \frac{\delta(t - \tau)}{|v_z|}
 \end{aligned} \quad (1)$$

with

$$\begin{aligned}
 \tau(z; \mathbf{p}'_{\perp}, \mathbf{x}'_{\perp}, t_o) = & t_o + \int_0^z \frac{1}{v_z(z'; \mathbf{p}'_{\perp}, \mathbf{x}'_{\perp}, t_o)} dz' \\
 \Omega_4 \equiv & \left(\frac{x_o}{\Xi_o} \right)^2 + \left(\frac{y_o}{\Psi_o} \right)^2 + \left(\frac{p_{x0}}{\Pi_{x0}} \right)^2 \\
 & + \left(\frac{p_{y0}}{\Pi_{y0}} \right)^2 \leq 1
 \end{aligned}$$

where Ω_4 denotes the hypervolume in phase space with boundaries Ξ_o , Ψ_o , Π_{x0} , and Π_{y0} ; T is the average transit time of an electron through the interaction region; $D(\mathbf{p}'_{\perp}, \mathbf{x}'_{\perp})$ is the distribution function over the hypervolume Ω_4 ; and $\sigma_{\parallel}(t_o)$ is the entrance-time distribution. The notation τ is the time when an electron, entering at t_o , arrives at the axial position z . The distribution functions are subject to the following normalizations:

$$\begin{aligned}
 \iiint_{\Omega_4} dp_{x0} dp_{y0} dx_o dy_o D(\mathbf{p}'_{\perp}, \mathbf{x}'_{\perp}) = & 1 \\
 \int_{-T/2}^{T/2} dt_o \sigma_{\parallel}(t_o) = & T.
 \end{aligned}$$

The hypervolume Ω_4 has the shape of a hyper-ellipsoid, and its volume can be used as a figure-of-merit of beam quality, such as the normalized four-volume emittance $\hat{\epsilon}_4 = V_4 = \frac{1}{2} \Xi_o \Psi_o \Pi_{x0} \Pi_{y0}$ [16] and the brightness $B_N = I_b / V_4$ [17]. Furthermore, $\hat{\epsilon}_4$ can be written as the product of x -plane and y -plane normalized root-mean-square emittances, i.e., $\hat{\epsilon}_4 = \frac{1}{2} \hat{\epsilon}_{x(\text{rms})} \hat{\epsilon}_{y(\text{rms})}$, where the root-mean-square emittance is defined by Fraser *et al.* [18]. Therefore we obtain the following results. For a Gaussian distribution,

$$\begin{aligned}
 \hat{\epsilon}_{x(\text{rms})} = & 2\pi \Xi_o \Pi_{x0}, \quad \hat{\epsilon}_{y(\text{rms})} = 2\pi \Psi_o \Pi_{y0} \\
 B_N = & 8 \frac{I_b}{\hat{\epsilon}_{x(\text{rms})} \hat{\epsilon}_{y(\text{rms})}}.
 \end{aligned} \quad (2a)$$

For a uniform distribution,

$$\begin{aligned}
 \hat{\epsilon}_{x(\text{rms})} = & \frac{2}{3} \pi \Xi_o \Pi_{x0}, \quad \hat{\epsilon}_{y(\text{rms})} = \frac{2}{3} \pi \Psi_o \Pi_{y0} \\
 B_N = & \frac{8}{9} \frac{I_b}{\hat{\epsilon}_{x(\text{rms})} \hat{\epsilon}_{y(\text{rms})}}.
 \end{aligned} \quad (2b)$$

Note that the same emittance will give a factor of nine difference in the value of brightness for the two distributions considered.

The wiggler field is generated by an array of parabolically shaped permanent magnets [19] and is shown to have a second-order focusing effect such that it enhances the FEL performance over the conventional linear dipole. We allow variation of $B_w(z)$ along the z axis for the study of efficiency enhancement. According to the KMR paper [11], the taper rate satisfying the equation

$$\frac{da_w}{dz} = -2k_w a_{ln}(z) \sin \Psi_{\text{res}}$$

ensures that the ln th ponderomotive wave stays resonant with the slowing electron beam, where a_w and a_{ln} represent the normalized vector potential of the wiggler and the ln th waveguide mode field, respectively. In addition, the KMR nonlinear wiggler taper is able to reduce unwanted competing modes, since each competing wave propagates by its own phase velocity, so that each ponderomotive wave has its own optimal taper rate.

If we assume that the space-charge force is weak compared with the ponderomotive force, the radiation field can be expanded in terms of the vacuum waveguide modes; i.e.,

$$\begin{aligned}
 A(\mathbf{x}, t) = & \sum_{\text{TE}_{ln}} A_{ln}^{\text{TE}}(z) \cos \alpha_{ln}(z, t) \hat{\mathbf{e}}_{ln}^1(x, y) \\
 & + \sum_{\text{TM}_{ln}} A_{ln}^{\text{TM}}(z) (\cos \alpha_{ln} \hat{\mathbf{e}}_{ln}^2(x, y) \\
 & + \sin \alpha_{ln} \hat{\mathbf{e}}_{ln}^3(x, y)).
 \end{aligned} \quad (3)$$

The notation

$$\alpha_{ln}(z, t) \equiv \left(\int_0^z k_{ln}(z') dz' - \omega t \right)$$

represents the phase of waveguide mode, and $\hat{e}_{\text{in}}^1(x, y)$, $\hat{e}_{\text{in}}^2(x, y)$ and $\hat{e}_{\text{in}}^3(x, y)$ are mode functions, defined as follows:

$$\begin{aligned}\hat{e}_{\text{in}}^1(x, y) &= \frac{ky}{k_{c,\text{in}}} \cos k_x \left(x + \frac{a}{2} \right) \sin k_y \left(y + \frac{b}{2} \right) \hat{x} \\ &\quad - \frac{k_x}{k_{c,\text{in}}} \sin k_x \left(x + \frac{a}{2} \right) \cos k_y \left(y + \frac{b}{2} \right) \hat{y} \\ \hat{e}_{\text{in}}^2(x, y) &= \frac{kx}{k_{c,\text{in}}} \cos k_x \left(x + \frac{a}{2} \right) \sin k_y \left(y + \frac{b}{2} \right) \hat{x} \\ &\quad + \frac{k_y}{k_{c,\text{in}}} \sin k_x \left(x + \frac{a}{2} \right) \cos k_y \left(y + \frac{b}{2} \right) \hat{y} \\ \hat{e}_{\text{in}}^3(x, y) &= \frac{k_{c,\text{in}}}{k_{\text{in}}} \sin k_x \left(x + \frac{a}{2} \right) \sin k_y \left(y + \frac{b}{2} \right) \hat{z} \quad (4)\end{aligned}$$

where

$$k_x = \frac{l\pi}{a}, \quad k_y = \frac{n\pi}{b}, \quad k_{c,\text{in}}^2 = k_x^2 + k_y^2$$

(a and b are the width and height of a rectangular waveguide), and $k_{\text{in}}(z)$ is the wave number of the l th mode.

We now can solve the inhomogeneous wave equation by substituting the expression of the vector potential $\mathbf{A}(x, t)$ and source current density $\mathbf{J}(x, t)$ and by utilizing the orthonormality properties. As a result, we obtain the following set of equations, which are identical to those derived in the Freund *et al.* paper [6] for the TE mode:

$$\begin{aligned}\frac{d^2}{dz^2} a_{\text{in}} + \left(\frac{\omega^2}{c^2} - k_{\text{in}}^2 - k_{c,\text{in}}^2 \right) a_{\text{in}} \\ = 2 \frac{\omega_b^2}{c^2} H_{\text{in}} \left\langle \hat{e}_{\text{in}}^1(x, y) \cdot \frac{\mathbf{v}}{|v_z|} \cos \alpha_{\text{in}} \right\rangle \\ 2\sqrt{k_{\text{in}}} \frac{d}{dz} (\sqrt{k_{\text{in}}} a_{\text{in}}) \\ = -2 \frac{\omega_b^2}{c^2} H_{\text{in}} \left\langle \hat{e}_{\text{in}}^1(x, y) \cdot \frac{\mathbf{v}}{|v_z|} \sin \alpha_{\text{in}} \right\rangle. \quad (5)\end{aligned}$$

For the TM mode,

$$\begin{aligned}\frac{d^2}{dz^2} a_{\text{in}} + \left(1 + \frac{k_{c,\text{in}}^2}{k_{\text{in}}^2} \right) \left(\frac{\omega^2}{c^2} - k_{\text{in}}^2 - k_{c,\text{in}}^2 \right) a_{\text{in}} \\ = 2 \frac{\omega_b^2}{c^2} H_{\text{in}} \\ \times \hat{e}_{\text{in}}^2(x, y) \cdot \frac{\mathbf{v}}{|v_z|} \cos \alpha_{\text{in}} + \hat{e}_{\text{in}}^3(x, y) \\ \cdot \frac{\mathbf{v}}{|v_z|} \sin \alpha_{\text{in}}\end{aligned}$$

$$\begin{aligned}2 \sqrt{k_{\text{in}}} \left(1 + \frac{k_{c,\text{in}}^2}{k_{\text{in}}^2} \right) \frac{d}{dz} \left(\sqrt{k_{\text{in}}} \left(1 + \frac{k_{c,\text{in}}^2}{k_{\text{in}}^2} \right) a_{\text{in}} \right) \\ = -2 \frac{\omega_b^2}{c^2} H_{\text{in}} \\ \times \hat{e}_{\text{in}}^2(x, y) \cdot \frac{\mathbf{v}}{|v_z|} \sin \alpha_{\text{in}} - \hat{e}_{\text{in}}^3(x, y) \\ \cdot \frac{\mathbf{v}}{|v_z|} \cos \alpha_{\text{in}} \quad (6)\end{aligned}$$

where

$$a_{\text{in}} \left(\equiv \frac{eA_{\text{in}}}{m_o c^2} \right)$$

is the normalized mode potential,

$$\omega_b \left(\equiv \sqrt{\frac{4\pi e^2 n_b}{m_o}} \right)$$

is the nonrelativistic plasma frequency, and

$$H_{\text{in}} = \begin{cases} 2, & l = 0 \text{ or } n = 0 \\ 4, & \text{otherwise.} \end{cases}$$

The average operator $\langle \cdots \rangle$ is defined as integration over the hypervolume Ω_4 and a full period of ponderomotive phase:

$$\begin{aligned}\langle \cdots \rangle \equiv \iiint_{\Omega_4} dp_{x_0} dp_{y_0} dx_0 dy_0 D(p_{\perp}^0, \mathbf{x}_{\perp}^0) \\ \cdot \int_{-\pi}^{\pi} d\psi_0 \sigma_{\text{II}}(\psi_0) (\cdots). \quad (7)\end{aligned}$$

The effect of electron beam source distribution is considered to be incorporated in the average operator.

To describe the electrons' motion, we need to solve the relativistic equations of motion for each electron, including the forces from the wiggler and waveguide fields (indexed by m), i.e.,

$$\begin{aligned}\frac{d\mathbf{p}_i}{dt} = -e \sum_m \left(\mathbf{E}_m + \frac{\mathbf{v}_i}{c} \times \mathbf{B}_m \right) - e \frac{\mathbf{v}_i}{c} \times \mathbf{B}_w \\ i = 1, 2, \cdots, N_{\text{part}}. \quad (8)\end{aligned}$$

Eventually, we obtain the following equations (omitting the index i for clarity):

$$\begin{aligned}v_z \frac{d}{dz} p_x = \frac{1}{\gamma} k_w c \sqrt{2} a_w [(\sqrt{2} W_3 \sin k_w z) p_y \\ + (W_2 \cos k_w z) p_z] \\ + \sum_{\text{TE}_{\text{in}}} m_o c a_{\text{in}} \left[(\omega - k_{\text{in}} v_z) \frac{k_y}{k_{c,\text{in}}} B_2 \sin \alpha_{\text{in}} \right. \\ \left. + v_y k_{c,\text{in}} B_3 \cos \alpha_{\text{in}} + \Gamma_{\text{in}} v_z \frac{k_y}{k_{c,\text{in}}} B_2 \cos \alpha_{\text{in}} \right]\end{aligned}$$

$$v_z \frac{d}{dz} p_x = + \sum_{\text{TM}_{ln}} m_o c a_{ln} \left\{ \left[\omega - k_{ln} \left(1 + \frac{k_{c,ln}^2}{k_{ln}^2} \right) v_z \right] \cdot \frac{k_x}{k_{c,ln}} B_2 \sin \alpha_{ln} + \Gamma_{ln} v_z \frac{k_x}{k_{c,ln}} B_2 \cos \alpha_{ln} \right\} \quad (9)$$

$$v_z \frac{d}{dz} p_y = - \frac{1}{\gamma} k_w c \sqrt{2} a_w [(\sqrt{2} W_3 \sin k_w z) p_x + (W_1 \cos k_w z) p_z] + \sum_{\text{TE}_{ln}} m_o c a_{ln} \left[(\omega - k_{ln} v_z) \frac{k_x}{k_{c,ln}} B_1 \sin \alpha_{ln} + v_x k_{c,ln} B_3 \cos \alpha_{ln} + \Gamma_{ln} v_z \frac{k_x}{k_{c,ln}} B_1 \cos \alpha_{ln} \right] + \sum_{\text{TM}_{ln}} m_o c a_{ln} \left\{ \left[\omega - k_{ln} \left(1 + \frac{k_{c,ln}^2}{k_{ln}^2} \right) v_z \right] \cdot \frac{k_y}{k_{c,ln}} B_1 \sin \alpha_{ln} + \Gamma_{ln} v_z \frac{k_y}{k_{c,ln}} B_1 \cos \alpha_{ln} \right\} \quad (10)$$

$$v_z \frac{d}{dz} p_z = - \frac{1}{\gamma} k_w c \sqrt{2} a_w [W_2 p_x - W_1 p_y] \cos k_w z + \sum_{\text{TE}_{ln}} m_o c a_{ln} (v_x k_y B_2 - v_y k_x B_1) \cdot \left(\frac{k_{ln}}{k_{c,ln}} \sin \alpha_{ln} - \frac{\Gamma_{ln}}{k_{c,ln}} \cos \alpha_{ln} \right) - \sum_{\text{TM}_{ln}} m_o c a_{ln} \left\{ \omega \frac{k_{c,ln}}{k_{ln}} B_4 \cos \alpha_{ln} - (v_x k_x B_2 + v_y k_y B_1) \times \left[\frac{k_{ln}}{k_{c,ln}} \left(1 + \frac{k_{c,ln}^2}{k_{ln}^2} \right) \sin \alpha_{ln} - \frac{\Gamma_{ln}}{k_{c,ln}} \cos \alpha_{ln} \right] \right\} \quad (11)$$

where

$$a_w(z) \left(\equiv \frac{e B_w(z)}{\sqrt{2} m_o c^2 k_w} \right)$$

is the normalized wiggler potential;

$$\Gamma_{ln} \left(\equiv \frac{d}{dz} \log_e a_{ln}(z) \right)$$

denotes the growth rate of the respective waveguide modes; $B_1(x, y)$, $B_2(x, y)$, $B_3(x, y)$, $B_4(x, y)$ are functions related to the transverse variation of the waveguide modes; and $W_1(x, y)$, $W_2(x, y)$, and $W_3(x, y)$ describe

the transverse dependence of the wiggler field as follows:

$$\begin{aligned} B_1(x, y) &= \sin k_x \left(x + \frac{a}{2} \right) \cos k_y \left(y + \frac{b}{2} \right) \\ B_2(x, y) &= \cos k_x \left(x + \frac{a}{2} \right) \sin k_y \left(y + \frac{b}{2} \right) \\ B_3(x, y) &= \cos k_x \left(x + \frac{a}{2} \right) \cos k_y \left(y + \frac{b}{2} \right) \\ B_4(x, y) &= \sin k_x \left(x + \frac{a}{2} \right) \sin k_y \left(y + \frac{b}{2} \right) \\ W_1(x, y) &= \sinh \left(\frac{k_w x}{\sqrt{2}} \right) \sinh \left(\frac{k_w y}{\sqrt{2}} \right) \\ W_2(x, y) &= \cosh \left(\frac{k_w x}{\sqrt{2}} \right) \cosh \left(\frac{k_w y}{\sqrt{2}} \right) \\ W_3(x, y) &= \cosh \left(\frac{k_w x}{\sqrt{2}} \right) \sinh \left(\frac{k_w y}{\sqrt{2}} \right). \end{aligned} \quad (12)$$

We then expand the equation $\mathbf{v} = d\mathbf{x}/dt$, and write the ponderomotive phase evolution for $\Psi_{ln}(z)$ for the i th electron with respect to the ln th waveguide mode. Equations (5), (6), and (9)–(11) constitute a set of coupled differential equations, which describes the wave-particle interaction in a 3-D manner and includes effects of mode competition, wiggler tapering, and beam source distributions.

III. SIMULATION RESULTS AT 280 GHz

We have developed a 3-D simulation code that consists of a beam-wave loader, a wave-particle solver, and a post-processor. In the beam-wave loader, we launch the waveguide modes with given power levels and prepare electrons with assigned beam brightness and distribution functions. Since we use the N th-order Gaussian quadrature technique to evaluate the average operator, electrons are loaded with coordinates and momenta specified by the technique itself. In the wave-particle solver, the fourth-order Runge-Kutta-Gill method is applied for solving the set of coupled differential equations. The numerical error for this algorithm is proportional to the fifth power of the step size. The post-processor then documents the simulation results with optional graphics.

Our version of the 3-D code has two different algorithms as compared with those of the Freund *et al.* 3-D code [6], [7]. First, the electron beam source in [6] and [7] has uncorrelated momenta and coordinate distribution functions, which is most suitable for the thermionic electron sources. However, for the induction linacs considered in this paper, as well as RF-linacs, the distributions of electron-particle momenta and coordinates are correlated and are more appropriately modeled by an aggregate distribution function over momenta and coordinate space. Second, our 3-D code allows the wiggler amplitude profile to be tapered nonlinearly as well as the linear taper that is considered in the previous works [6], [7].

Our 3-D code also differs when compared with the wiggler-averaged WFEL code [20]. First of all, the WFEL code simulates only the slow time-scale mechanisms for the average over a wiggler period, eliminating the fast timescale variation; our 3-D code keeps both timescale variations. Second, the inclusion of mode competition in the WFEL code is via the calculation of mode-coupling coefficients with explicit analytic expressions [20]; our 3-D code includes mode competition implicitly through (5)–(6) and the equations of motion (9)–(12). The electric space-charge field in the WFEL code is calculated analytically by treating the electron beam as a wiggling fluid flow and solving the wave equation subject to modified boundary conditions. To simplify the algebraic calculation, the size of electron beam is assumed to be much smaller than the waveguide size, and the average process with the FEL synchronism condition satisfied is utilized.

A simulation of an FEL amplifier operating at 280 GHz has been carried out with the following parameters: $V_b = 10$ MeV, $I_b = 3000$ A, $r_b = 0.4537$ cm, $\epsilon = 251$ mrd · mm, $B_N = 1 \times 10^4$ A/(rad · cm)² with $B_w = 4.09$ kG, $\lambda_w = 10$ cm, and a guide which is 3.10 cm². The TE₀₁ mode is injected at 458 W with residual TM₂₁, TE₂₁, and TE₀₃ modes excited at 65-mW levels, corresponding to the anticipated values for the planned Livermore FEL experiment [2]. This FEL is operated in the transition between the Compton and the Raman regimes, or the marginal Raman regime, when a comparison of ponderomotive and space-charge forces and length of the wiggler are considered. For the uniform wiggler case, as shown in Fig. 2, the total saturation power is 1.4 GW at $k_w z = 142$ ($z = 2.16$ m), which corresponds to a growth rate 30 dB/m or an efficiency of 4.6 percent. The TM₂₁ and TE₂₁ modes are shown to compete substantially with the TE₀₁ mode, reaching 50 percent of the wave power near $k_w z = 50$.

We then add a nonlinear taper section for efficiency enhancement. As pointed out by Jong *et al.* [3], the optimal beginning taper position is not solely determined by maximizing the TE₀₁ power at the end of the constant section but, more importantly, by considering energy and phase spread within the ponderomotive potential well. This is because significant energy spread will prevent electrons from remaining trapped in the ponderomotive well. Fig. 3 shows electron evolution in (γ, ψ) space at different axial positions and indicates that $k_w z = 115$ is the optimal tapering position according to preceding argument. On the other hand, the beam confinement by the wiggler is proportional to the wiggler's magnitude, so that we terminate the tapering before the electron beam blows up and intersects the wall. As a result, the efficiency is enhanced to 48 percent and the higher-order modes considered are reduced to less than 2 percent, as shown in Fig. 4. In other words, there is 14.1 GW in the TE₀₁ mode together with 0.3 GW in higher-order modes at the end of wiggler. We also examined a linear taper with the rate $da_w/dz = -0.10$ cm⁻¹ and found that higher-order modes contain up to 10 percent of the total wave power with the effi-

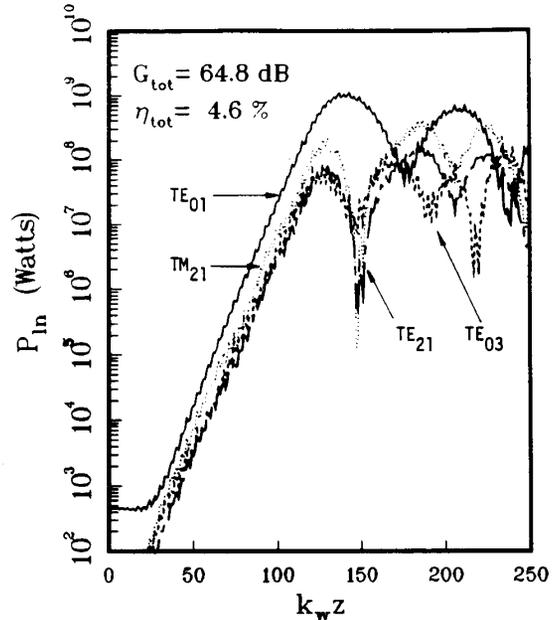


Fig. 2. Solution of 280-GHz mode power for untapered wiggler.

ciency reduced to 42 percent, as compared with a total higher-order mode power of 2 percent with an efficiency of 48 percent for the nonlinear taper case. This shows the superiority of the KMR nonlinear taper to the linear taper for efficiency enhancement and higher-order mode suppression. We also looked at the sensitivity of the TE₀₁ power to the assumed higher-order-mode input power levels. We found that, at higher power levels, they can influence the TE₀₁ power evolution, as is evident in Fig. 5. Note that the TE₀₁ mode damps initially and then grows normally when higher-order mode power is increased up to 65 W (three orders of magnitude). Nevertheless, the TE₀₁ power is reduced by only 6 percent at the end of the wiggler (cf. the case for Fig. 4).

As mentioned in the preceding, we neglect space-charge forces in the 3-D code. To address the effects of the space-charge force, we compare the results with those from the WFEL code. The WFEL code, developed by Byers and Cohen [20], is a wiggler-averaged simulation that includes space-charge forces. It has been shown to have excellent agreement with the 35-GHz FEL experiment at the Lawrence Livermore National Laboratory. In Fig. 6, our 3-D results, compared with those from the WFEL code without space-charge forces, show that the growth rate during the initial constant wiggler section and in the taper region are slightly different. Due to this difference, and the constraint that the taper must continue down to a field of 600 G, the total interaction lengths differ by 10 percent (3.6 m for the 3-D code versus 4.0 m for the WFEL code). This indicates that there are some slight differences between the 3-D and wiggler-averaged formulations. If space-charge forces are included in the WFEL code, we find that the growth rate both before and within

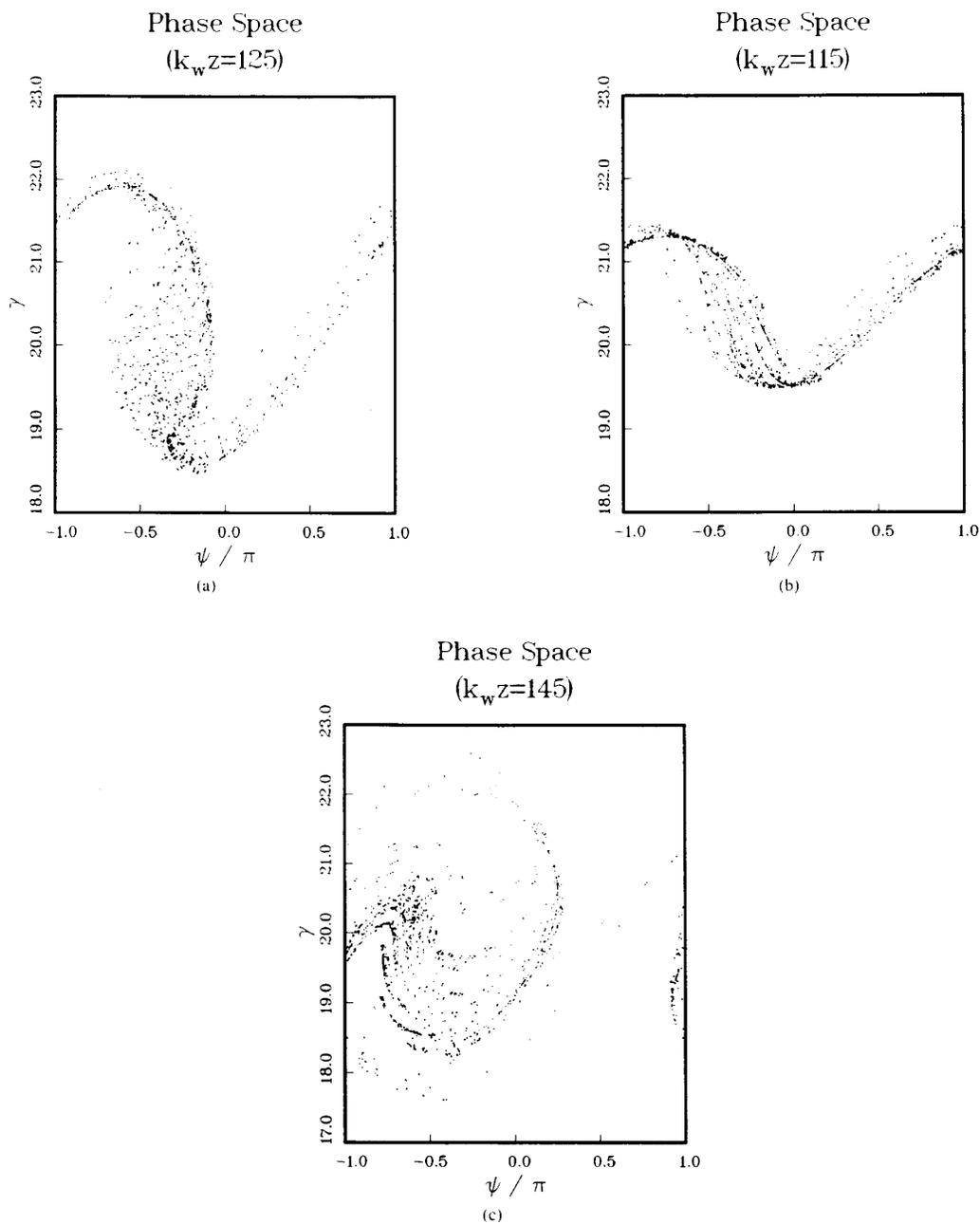


Fig. 3. Phase space (γ , ψ) evolution of electron beam. (a) At $k_w z = 115$. (b) At $k_w z = 125$. (c) At $k_w z = 145$.

the taper section is significantly reduced. This results in reducing the efficiency to 32 percent as well as changing the wiggler taper profile.

To compare our works with others, we examine the results discussed in the paper by Ganguly and Freund [10]. The closest they come to the case discussed in this paper is a 1-MeV 50-A case for which the efficiency reduction due to space charge is typically under 5 percent at a frequency near 35 GHz. The addition of the strong guide

field, shorter total wiggler length, lower beam current, and lower operating frequency for their case is the reason for a 5-percent reduction in efficiency as compared to the 16-percent reduction in efficiency for our case.

We also compare the results of Fig. 6 with those obtained using the FRED code discussed in [2] and [3]. Comparison with [2, fig. 6] and [3, fig. 4] shows efficiencies from 37 to 33 percent for 6 m ($B = 10^5 \text{ A}/(\text{rad-cm})^2$) and 4.6 m ($B = 10^4 \text{ A}/(\text{rad-cm})^2$), respectively. For our

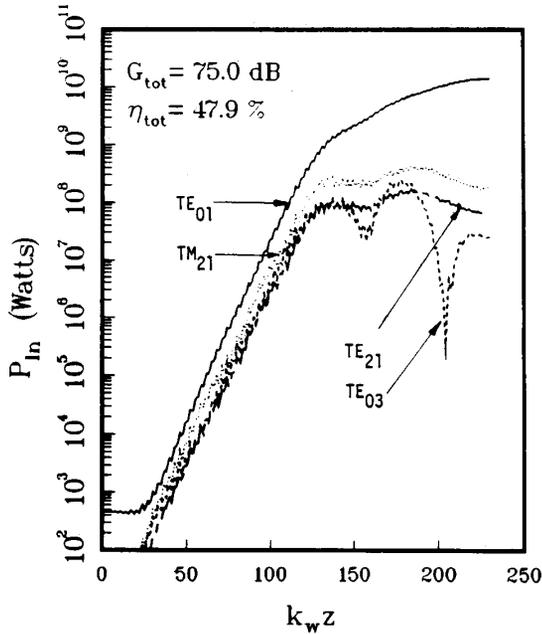


Fig. 4. Evolution of 280-GHz mode power for nonlinear taper.

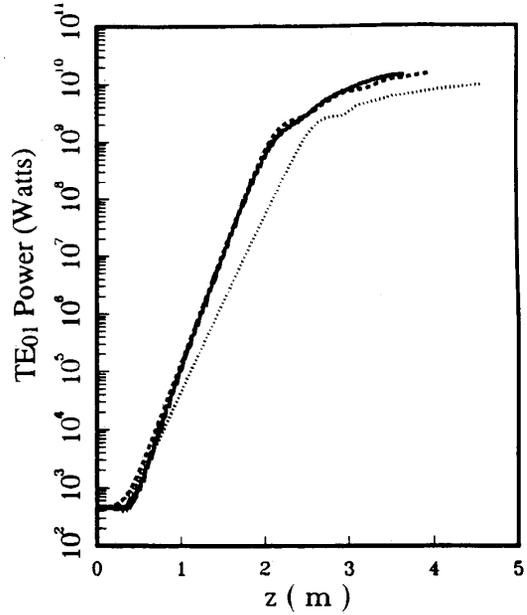


Fig. 6. Comparison of 3-D solution (solid line) and solutions from wiggler-averaged code without (dashed line) and with (dotted line) space-charge forces.

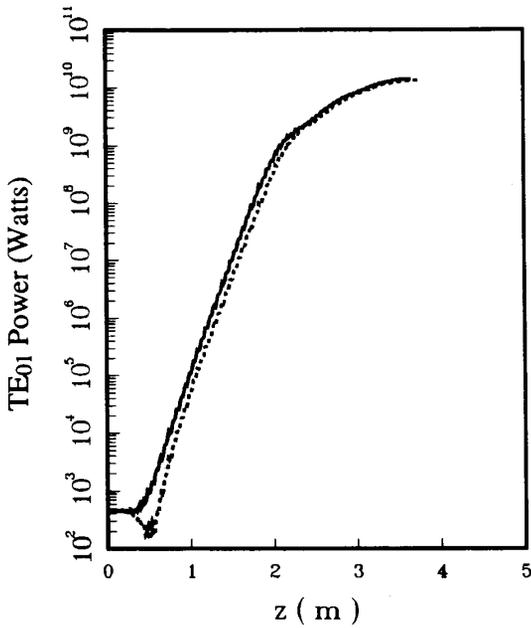


Fig. 5. Evolution of TE₀₁ power with higher-order mode input power at 68 mW each (solid line) and at 68 W each (dashed line).

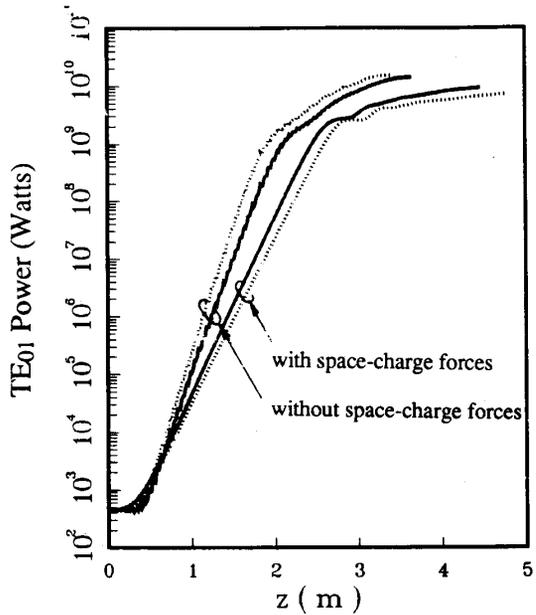


Fig. 7. Comparison of uniform (solid line) and Gaussian (dotted line) source distributions without and with space-charge forces.

comparable case, the efficiency is 32 percent for a wiggler of 4-m length with a beam brightness of 10^4 A/(rad · cm)². Although the exact same cases were not run for both codes, the code results appear comparable. Our results yield a higher efficiency per unit wiggler length due to our use of a 0.55-rad psi resonant (which determines

the taper properties) rather than that of 0.1–0.2 rad, which was used for their cases.

We also investigated the effect of electron beams with different source distributions over the regimes from small signal to saturation in the millimeter wave regime. Two cases are examined: one is uniform distribution and the

other is Gaussian distribution. First, we look at the condition with the same brightness: $1 \times 10^4 \text{ A}/(\text{rad} \cdot \text{cm})^2$ (emittance is changed following (2)). We find that, in both cases, the growth rate and the power evolution are almost the same. Second, we fix the emittance at $251 \text{ mrd} \cdot \text{mm}$ and change the brightness according to (2), which makes the brightness $1 \times 10^4 \text{ A}/(\text{rad} \cdot \text{cm})^2$ for the uniform distribution and $9 \times 10^4 \text{ A}/(\text{rad} \cdot \text{cm})^2$ for the Gaussian distribution. Fig. 7 shows that, without the inclusion of space-charge forces, the growth rate of the Gaussian distribution is larger than that of the uniform. In addition, the 15.3 GW of TE_{01} output power for the Gaussian distribution is 8.5-percent larger than the 14.1 GW for the uniform case. Note that different optimal taper positions are chosen for each case. If space-charge forces are included, the TE_{01} output power of the Gaussian distribution is reduced even further compared to that of the uniform case. Qualitatively, the brighter beam allows a smaller beam radius to carry the same amount of current, so that space-charge forces become stronger and reduce the efficiency further.

IV. CONCLUSION

We have carried out a 3-D simulation of an FEL amplifier at 280 GHz, including effects of wiggler tapering, mode competition, and different beam source distributions. The nonlinear wiggler taper has been shown to be superior to the linear variety in terms of efficiency enhancement and unwanted mode suppression. Our 3-D results agree very well with those from the wiggler averaged code when space-charge effects are neglected; moreover, we find space-charge forces reduce the efficiency from 48 percent down to a level of 32 percent for a 10-MeV 3-kA beam and 10-cm period linear dipole wiggler.

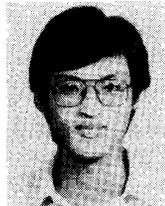
The sensitivity of FEL gain to higher-order mode power and electron beam source distributions is examined as well. We infer that higher-order-mode power levels varying over three orders of magnitude can affect the TE_{01} power evolution but only slightly reduce the TE_{01} output power. As to the effect of beam source distributions, we find that only distributions with same emittance have observable differences. In the Compton regime, an electron beam with Gaussian distribution gives a higher growth rate and higher TE_{01} output power; however, the uniform distribution has larger growth rate and TE_{01} output power if space-charge forces are included.

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